

Integrated Memory Having A Voltage Generator Circuit For Generating A Voltage Supply For A Read/Write Amplifier

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CROSS REFERENCE TO RELATED APPLICATIONS

This application claims priority under 35 USC §119 to German Application No. 10316581.9, filed on April 10, 2003, and titled "Integrated Memory Having a Voltage Generator Circuit for Generating a Voltage Supply for a Read/Write Amplifier," the
10 entire contents of which are hereby incorporated by reference.

FIELD OF THE INVENTION

The present invention relates to an integrated memory having a memory cell array, which has word lines for selecting memory cells and bit lines for reading out or
15 writing data signals, and also having a read/write amplifier, which is connected to the bit lines for the purpose of assessing and amplifying data signals.

BACKGROUND

An integrated memory, for example, in the form of a DRAM, generally has a
20 memory cell array comprising word lines and bit lines. In this case, the memory cells are arranged at crossover points of the bit lines and word lines. The memory cells are constructed, in particular, from a storage capacitor and a selection transistor, the selection transistor connecting the respective storage capacitor to one of the bit lines. The control terminals of the respective selection transistors are respectively connected to one of the
25 word lines for the purpose of selecting the memory cells. An activated word line respectively turns on connected selection transistors. After the relevant word line has been selected, data signals of the memory cells along the selected word line are present

on the corresponding bit lines. A data signal of a selected memory cell is assessed and amplified in a read/write amplifier of the memory cell array. During a read access, the data signals of selected memory cells are read out for further processing and, during a write access, data signals to be written are written to the selected memory cells.

5 During a memory access, a word line is first of all activated. As a result, the memory cells arranged along a word line are respectively connected up conductively to a bit line via the relevant selection transistor. In this case, the stored charge is divided up in accordance with the memory cell capacitance and bit line capacitance. In accordance with the ratio of these two capacitances, i.e., a transfer ratio, this leads to deflection of the
10 bit line voltage. The read/write amplifier situated at one end of the bit line can assess this voltage and can amplify the relatively low potential difference until the bit line has reached the full signal level for a stored logic 1 (corresponding, for example, to a positive supply potential) or the signal level for a logic 0 (corresponding, for example, to a reference potential). These full signal levels are provided by a voltage generator circuit,
15 which is connected to the relevant read/write amplifier.

 The magnitude of the supply voltage of memories is being constantly reduced, particularly in view of increasing demands for reliability and low energy consumption. In the course of the reduction, modern integrated memories regulate an externally applied supply voltage to a smaller supply voltage within the memory. On the other hand, higher
20 processing speeds of memories and higher data throughput are demanded, particularly on account of increasing memory size. However, it holds true, particularly with regard to the voltage supply for a read/write amplifier of an integrated memory, that a lower supply voltage for reducing the power consumption also leads to a reduction in the switching

speed of the relevant read/write amplifier, if the read/write amplifier for the assessment and amplification operation is activated using the lower supply voltage.

SUMMARY

5 An integrated memory, in which an assessment and amplification operation can be carried out by a read/write amplifier at a comparatively high switching speed and in which a low power consumption can be provided.

 In accordance with the invention, the voltage generator circuit of the integrated memory of the type mentioned can generate a voltage supply for application to the
10 read/write amplifier. A potential difference can be applied to the read/write amplifier using different supply potentials. The voltage generator circuit can increase the potential difference applied to the read/write amplifier for a limited period of time during an assessment and amplification operation of the read/write amplifier. In accordance with the invention, an increased potential difference for application to the read/write amplifier
15 can be made available for that part of an assessment and amplification operation of the read/write amplifier, which is critical for the switching speed, with the result that comparatively high switching speeds of the read/write amplifier can be possible. At the same time, the power consumption of the integrated memory can be reduced as a result of an applied potential difference being comparatively low for the remainder of the period of
20 time.

 In accordance with one embodiment of the present invention, the voltage generator circuit can increase a first supply potential at a first terminal of the read/write amplifier and/or can lower a second supply potential at a second terminal of the

read/write amplifier during an assessment and amplification operation of the read/write amplifier.

In accordance with the invention, charge-dependent control to generate the increased potential difference using a defined quantity of charge can be implemented in the voltage generator circuit. This means that the increased potential difference can be applied to the read/write amplifier in a charge-controlled manner by a defined capacitance being charged, for example, before the relevant assessment and amplification operation, said capacitance then being discharged again during the assessment and amplification operation.

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BRIEF DESCRIPTION OF THE FIGURES

The invention is explained in more detail below with reference to the FIG. s which represent exemplary embodiments for the present invention and are illustrated in the drawing, in which:

15 FIG. 1 shows a diagrammatic illustration of a memory cell array of an integrated memory in accordance with the invention,

FIG. 2 shows a potential profile of a voltage supply for a read/write amplifier in accordance with the invention,

20 FIGS. 3A and 3B show an embodiment of a voltage generator circuit having a read/write amplifier, with associated signal diagram, and

FIGS. 4A and 4B show an embodiment of a voltage generator circuit having a read/write amplifier in accordance with the invention, with associated signal diagram.

DETAILED DESCRIPTION

FIG. 1 shows a diagrammatic illustration of a memory cell array of an integrated memory M, in which memory cells MC are arranged along word lines WL0, WL1, WL2 and bit lines BL0, BL1. The memory cells MC can be arranged at crossover points of the bit lines BL0, BL1 and word lines WL0, WL1, WL2. In the present exemplary embodiment, a limited number of word lines and bit lines are shown for the sake of clarity but in practice, an integrated memory has a plurality of word lines and bit lines. The memory cells MC can each include a storage capacitor C, which is connected to one of the bit lines BL0, BL1 via a selection transistor AT. In order to select one of the memory cells MC, the respective selection transistor AT can be turned on by an activated word line WL0, WL1, WL2 such that a data signal can subsequently be read out from or written to a selected memory cell. The data signal of the selected memory cell is present on the relevant bit line BL0, BL1 and is assessed and amplified in one of the illustrated read/write amplifiers SA0, SA1.

During an operation of reading out a data signal from one of the memory cells, the stored charge can be divided up in accordance with the memory cell capacitance and bit line capacitance. In accordance with the ratio of these two capacitances, this can lead to deflection of the bit line voltage. The read/write amplifiers SA0, SA1 can assess this bit line voltage and, when reading the memory cells, can amplify the relatively low potential difference until the relevant bit line has reached the full signal level for a stored logic 1 (corresponding, for example, to the positive supply potential V1) or the signal level for a logic 0 (corresponding, for example, to the supply potential V2). The supply potentials V1 and V2 can be generated by a voltage generator circuit (not illustrated in FIG. 1), with

the result that a corresponding potential difference is present at the sense amplifiers SA0, SA1.

FIG. 2 shows a potential profile of a voltage supply for the read/write amplifiers SA0, SA1 shown in FIG. 1, in accordance with the principle according to the invention.

5 At the beginning of an assessment and amplification operation, the supply potentials $V1 = V_{blh}$ and $V2 = \text{gnd} + V$ can be present at the read/write amplifiers SA0, SA1. This comparatively low potential difference between the supply potentials $V1$ and $V2$ can make possible a comparatively low power consumption of the memory. In order to increase the switching speed during the assessment and amplification operation of the
10 read/write amplifier, an increased potential difference can be applied to the relevant read/write amplifier SA0, SA1 shown in FIG. 1 for the period of time T between the instants $t1$ and $t2$, i.e., at $V1 = V_{blh} + V$ and $V2 = \text{gnd}$ in the example. In this case, therefore, the supply potential $V1$ at the upper terminal of the relevant read/write amplifier is increased and the supply potential $V2$ at the lower terminal of the relevant
15 read/write amplifier can be lowered during the assessment and amplification operation of the read/write amplifier SA0, SA1.

FIGS. 3A and 3B show an embodiment of a voltage generator circuit having a read/write amplifier SA, in which circuit time control is implemented, in accordance with which the potential difference is increased in a defined period of time. The voltage
20 generator circuit VG1 (FIG. 3A) can have a pulse shaper PF, which can drive a PFET transistor P1 and an NFET transistor N1. The transistors P1 and N1 can be connected to the supply potentials $V_{blh} + V$ and gnd , respectively. The two transistors P1, N1 can be driven with signals $/P$ and P , respectively, which can be inverted with respect to one another. In contrast, the PFET transistor P2 can be connected to the supply potential

V_{blh} and the NFET transistor N2 can be connected to the supply potential $\text{gnd} + V$. In accordance with the illustration shown in FIG. 3A, two supply paths SP1, SP2 of the voltage generator circuit VG1 can be provided for the read/write amplifier SA. The supply paths can have a different potential difference.

5 FIG. 3B illustrates a signal diagram, in accordance with which the voltage generator circuit VG1 shown in FIG. 3A is operated. For an assessment and amplification operation (starting at the instant t_1) of the sense amplifier SA, the control signal SET for the pulse shaper PF can be switched to the active state. As a result, the pulse shaper PF can generate an active signal P, which can turn on the transistor N1. The
10 transistor P1 can be turned on by the inverted signal \overline{P} . A control pulse P or \overline{P} having a defined time duration T can be generated at the beginning of the assessment and amplification operation of the read/write amplifier SA. The control pulse driving the supply path SP1 can have the higher potential difference. Before the end of the assessment and amplification operation, the control pulse P can be deactivated at the
15 instant t_2 and the control signals NSET and PSET are activated for driving the transistors N2 and P2, respectively, with the result that the lower potential difference of the second supply path SP2 can be present at the read/write amplifier SA.

 FIG. 4A shows an embodiment of a voltage generator circuit VG2 having a read/write amplifier SA in accordance with the invention. Charge-dependent control,
20 which can generate the increased potential difference at the sense amplifier SA using a defined quantity of charge, can be implemented in the voltage generator circuit VG2 shown in FIG. 4A. Like the embodiment shown in FIG. 3A, the voltage generator circuit VG2 shown in FIG. 4A can have two supply paths SP1, SP2 for the read/write amplifier SA. The supply paths can be operated with a different potential difference. Like the

corresponding supply paths shown in FIG. 3A, the supply paths SP1, SP2 shown in FIG. 4A can be driven alternatively to one another in time and can be connected to the read/write amplifier SA. The capacitances C1, C2 which can be connected to the supply potentials gnd and gnd + V, respectively, can be connected into the supply path SP1, which can be intended for providing a higher potential difference for the read/write amplifier SA. The capacitances C1, C2 may be connected to the read/write amplifier SA via the PFET transistor P4 and NFET transistor N4, respectively. The transistors P4 and N4 can be driven by the control pulses /P and P, respectively, of a pulse shaper PF. The PFET transistor P3 and the NFET transistor N3 which are connected to the supply potentials Vblh + V and gnd, respectively, are provided for the purpose of charging and discharging the capacitances C1, C2. The transistor P3 can be driven by the control pulse P and the transistor N3 can be driven by the inverted control pulse /P. The read/write amplifier SA may be connected to the supply potential Vblh via the PFET transistor P5 of the second supply path SP2 and to the supply potential gnd + V via the NFET transistor N5.

FIG. 4B shows a signal diagram for operating the voltage generator circuit VG2 shown in FIG. 4A. At the beginning of an assessment and amplification operation by the read/write amplifier SA, the control signal SET can be activated at the instant t1 for driving the pulse shaper PF. The capacitances C1, C2 can be connected, by the pulse signals P, /P, to the read/write amplifier SA for the period of time T and can be discharged (C1) and charged (C2) in this state. Before the end of the assessment and amplification operation by the read/write amplifier SA, the control pulse P can be deactivated at the instant t2 and the transistors P5 and N5 can be turned on by the control signals PSET, NSET. Furthermore, the transistors P3 and N3 can be turned on, with the

result that the capacitances C1, C2 can be charged and discharged, respectively, for a next assessment and amplification operation. In the case of a next assessment and amplification operation by the read/write amplifier SA, the capacitances can then be connected up at the beginning of the assessment and amplification operation by a new control pulse P, /P of the pulse shaper PF. As a result, the increased potential difference is provided from previously precharged capacitances, which can be isolated from the increased supply over a defined period of time. Since the charge of the capacitances C1, C2 can be reversed using only a relatively low potential difference, the power consumption of the read/write amplifier SA can be limited.

While the invention has been described in detail and with reference to specific embodiments thereof, it will be apparent to one skilled in the art that various changes and modifications can be made therein without departing from the spirit and scope thereof. Accordingly, it is intended that the present invention covers the modifications and variations of this invention provided they come within the scope of the appended claims and their equivalents.

LIST OF REFERENCE SYMBOLS

	M	Integrated memory
	WL0, WL1, WL2	Word line
5	BL0, BL1	Bit line
	MC	Memory cells
	AT	Selection transistor
	C	Storage capacitance
	SA0, SA1	Read/write amplifier
10	V1, V2	Supply potential
	Vblh, Vblh + V	Supply potential
	gnd, gnd + V	Supply potential
	t1, t2	Instant
	T	Time duration
15	VG1, VG2	Voltage generator circuit
	SP1, SP2	Supply path
	PF	Pulse shaper
	SA	Read/write amplifier
	P1 to P5	Transistor
20	N1 to N5	Transistor
	P, /P	Control pulse
	SET	Control signal
	NSET, PSET	Control signal
	C1, C2	Capacitance